

Analysis of a Single Shorted Rectangular Microstrip Antenna for 50Ω Microstrip Line Feed

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Abstract: In this paper, analysis of a lower order and dominant mode of a rectangular microstrip antenna with a loading of the single shorting post has been presented. Microstrip line feed of 50Ω impedance along one of the radiating edges with a single shorting post has been proposed for impedance matching. A detailed investigation of the rectangular microstrip antenna with and without single shorting pin has been presented and experimentally validated with good agreement.

Keywords - Fundamental mode, lower mode, microstrip antenna, shorting post.

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I. Introduction

In recent years, compact antenna has received much attention due to their smaller size, as conventional antennas have size too large that do not fit in wireless systems or personal communications systems. In addition, the attractive features of low cost, lightweight, low profile and ease of fabrication make shorted microstrip antennas (MSA) as a good choice over other antennas [1]. MSA can be easily integrated with active devices and microwave integrated circuits by properly choosing direct feeding techniques of microstrip line [12]. Input impedance and characteristics of the MSA are generally influenced by the feeding technique, so it is a crucial antenna design parameter. In order to make a total antenna structure to be planar, feeding arrangements such as edge fed and microstrip fed can be etched on the same substrate layer. A MSA has a high input impedance at a radiating edge, which causes difficulty in impedance matching. Detailed investigation to solve the input impedance of an edge fed MSA is carried out with the help of the cosine function in [3]. In an alternative method of inset fed, input impedance is sensitive to notch width and depth which lead to increase in the cross-polarization level in H-plane as the patch geometries become asymmetrical [4].

Compact and shorted configurations of rectangular, circular and triangular antennas with the shorting post/pin technique have been reported in the literature [1, 5-8]. Shorting plates or shorting posts along the zero field line of the fundamental mode help to realize compact microstrip antenna [7]. Recently, in [9] impedance agile rectangular microstrip antenna (RMSA) loaded with single shorting pin inside the patch is proposed. The shorting helps to reduce the input impedance of the radiating edge of MSA and facilitate a microstrip feed of characteristic impedance of 110Ω. Generally, microstrip line of 110Ω impedance is not used, instead a 50Ω feed line is often preferred so that it is easily integrated with microwave circuit. Furthermore, the shorting also introduces lower mode frequency besides altering dominant modes, which has not been discussed in the paper [9].

In this paper, detailed investigation of lower order mode and the dominant mode frequency of a RMSA with single shorted post and 50Ω microstrip line feed along the radiating edge is presented and results are validated using experiments.

II. Antenna Configuration

A RMSA has been designed at the fundamental mode frequency (TM₁₀=FF) of 1.56 GHz with dielectric substrate

R04003C having a relative permittivity of $\omega = 3.55$, a dielectric thickness of $h = 1.524$ mm and loss tangent $\tan\delta =$

0.0027. The length and width of $L=W=50$ mm are calculated using the formulations given in [1].

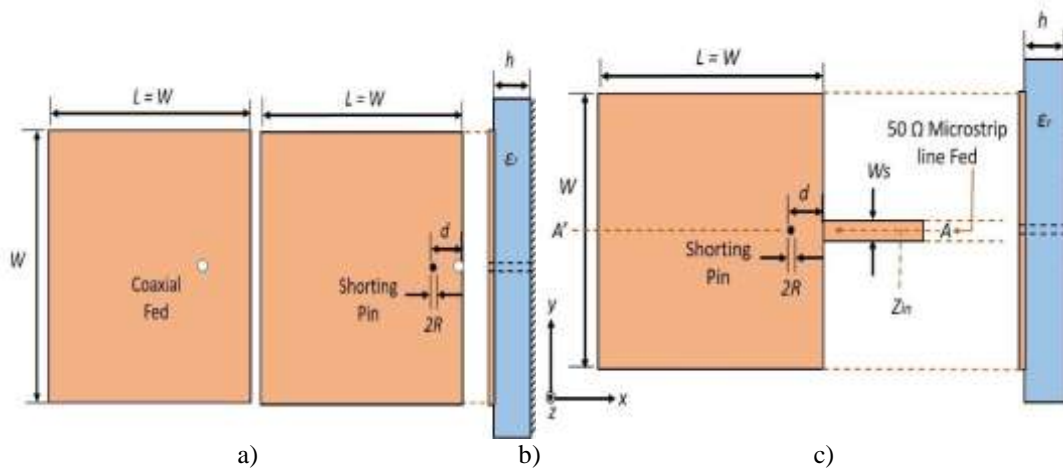


Fig. 1. Geometry of RMSA a) Coaxial fed, b) End Coaxial fed shorted and c) Shorted RMSA fed along radiating edge

III. Antenna Analysis

A. Modal analysis of basic co-axial fed RMSA

For the fundamental TM_{10} mode the center of the patch has zero potential. Thus, by putting a short at the center the characteristic of TM_{10} mode does not alter. However, it introduces the additional lower order mode.

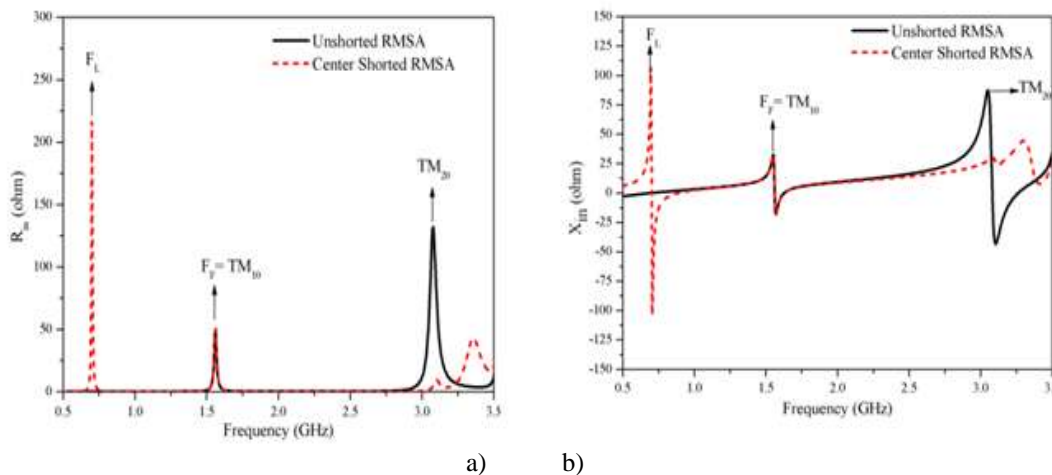


Fig. 2. Simulated results of co-axial fed RMSA with and without center shorting post (a) Input resistance and (b) Input reactance

Fig. 2(a) and (b) depict the simulated input resistance (R_{in}) and reactance (X_{in}) results of a coaxial fed RMSA with and without a center shorting post. The 50Ω co-axial feed position was determined as 7.35 mm from the center. It is confirmed from Fig. 2 that, if RMSA is shorted at center ($d/w=0.5$) via the shorting post of radius $R=0.5$ mm, produces the lower order mode frequency (F_L) while the fundamental mode frequency (F_F) remains unaltered. F_L corresponds to the condition that the length from shorting point to the opposite corner along the radiating edge of the RMSA is $\lambda/4$. There is a slight increase in higher order mode frequency (TM_{20}) due to the shorting post at center.

B. Investigation of shorting position versus F_L and F_F

An end coaxial fed RMSA with shorting posts for 50Ω match is shown in Fig. 2(b). To analyze the effect on F_L and F_F simulation has been performed with position of the shorting post d/w . Table I gives the variation of F_L and F_F of shorted RMSA with different shorting post position (d/w). Fig. 3 depicts the same effect graphically. It has been observed that with an increase in d/w from 0 to 0.5, F_F increases while F_L decreases. The feed point has to be located close to the shorting post for impedance matching. The maximum reduction in the F_F is obtained when the shorting post is placed at a radiating edge of the RMSA [1, 8]. For fundamental TM_{10} mode voltage distribution varies from zero at center short location ($d/w=0.5$) to a maximum value at the edge ($d/w=0.05$) of the RMSA. The resonance frequency F_L at center is approximately calculated by equating $L_e/2 + W_e/2 = \lambda/4$. Resonance frequency F_L formulation of single shorted RMSA is reported in [1, 8]. When shorting post along the center line moves from edge to the center, effective L_e (which is a length

from shorting post to the corner of the opposite radiating edge) decreases making $L_e + W_e/2$ shorter, thus the F_L decreases. Fig.4 compares simulated and calculated frequency (F_L) for different d/w ratio. The increase in F_F with increase in d/w is because of modification of field distribution caused by the shorting post and it does not remain pure TM₁₀ mode.

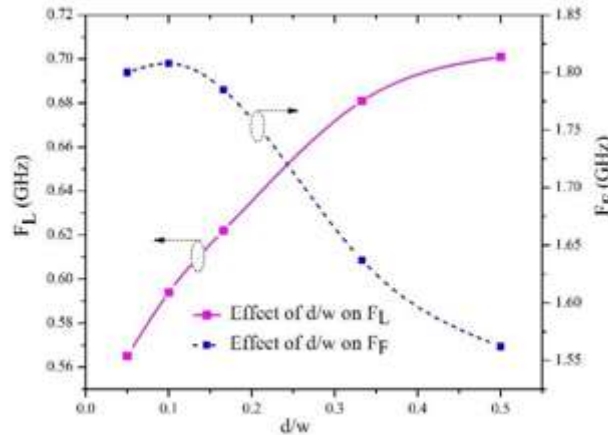


Fig. 3. Effect of position d/w of shorting position on F_L and F_F

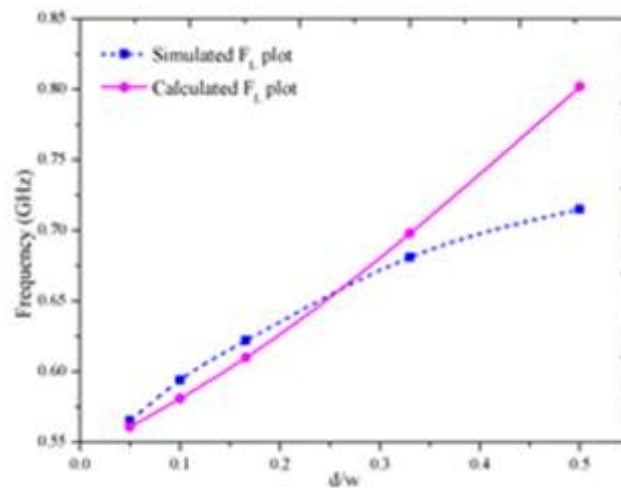


Fig.4. Comparison plot for simulated and calculated F_L with different d/w ratio

TABLE I. VARIATION OF F_L AND F_F WITH DIFFERENT SHORTING POSITION

| Sr.No. | Configuration (Freq.in GHz) | F_L | F_F |
|--------|----------------------------------|-------|--------|
| 1 | Un-shorted Simple RMSA | - | 1.560 |
| 2 | Center Short RMSA ($d/w=1/20$) | 0.701 | 1.562 |
| 3 | Shorted ($d/w=1/3$) | 0.681 | 1.6379 |
| 4 | Shorted ($d/w=1/6$) | 0.622 | 1.785 |
| 5 | Shorted ($d/w=1/10$) | 0.594 | 1.808 |
| 6 | Shorted ($d/w=1/20$) | 0.565 | 1.800 |

C. Single post loaded RMSA with 50Ω Microstripline feed at radiating edge

Fig.1c) depicts the 50Ω microstrip line feed along radiating edge of shorted RMSA. The microstripline dimensions are $W_s=3.402$ mm and $L_s=10$ mm with the single shorting post of radius $R=0.5$ mm, introduced along the center line AA'. Due to the shorting post, a surface current density at that location gets strengthened, which helps to reduce the edge impedance of the RMSA. Fig.5(a) and (b) depict the input resistance and reactance, respectively, of 50Ω microstripline

feed along radiating edge of shorted RMSA with different d/w ratio. It is observed that as d/w ratio increases (pin moves towards the center from radiating edge) i.e. $d/w=0.05$ to 0.5 , the input impedance of F_F and F_L reduced from 420Ω to 50Ω and 330Ω to 46Ω , respectively.

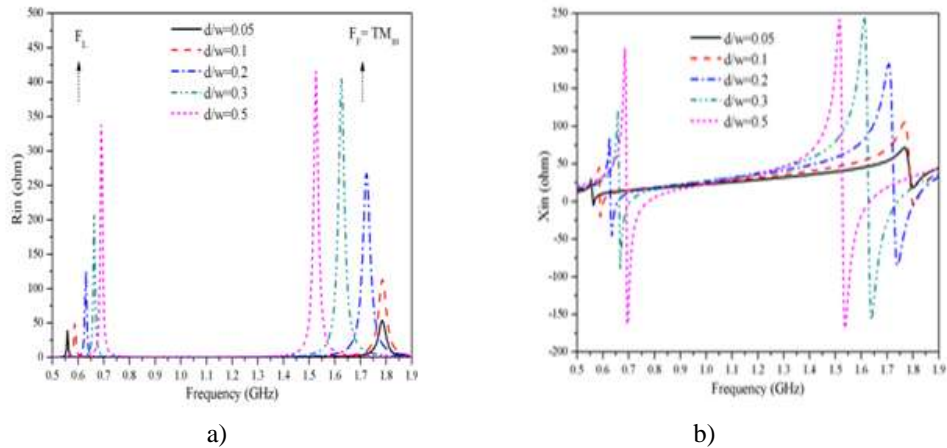


Fig. 5. Simulated results of 50Ω microstrip line fed shorted RMSA with different d/w ratio (a) Input resistance and (b) Input reactance

The resonance frequency of F_F is increased from 1.56GHz to 1.793GHz while F_L is decreased from 0.7GHz to 0.55GHz . Hence, by choosing the proper position of a shorting post wide range of input impedance of microstrip edge fed RMSA can be adjusted for various frequency. The proposed approach of 50Ω microstrip line feed RMSA eliminates the use of separate TRL calibration kit, which is proposed in [9] for de-embedding the real input impedance. Fig. 6 depicts the simulated co-polarization in E-plane and cross-polarization in H-plane of the 50Ω microstrip line edge fed RMSA with different d/w ratio.

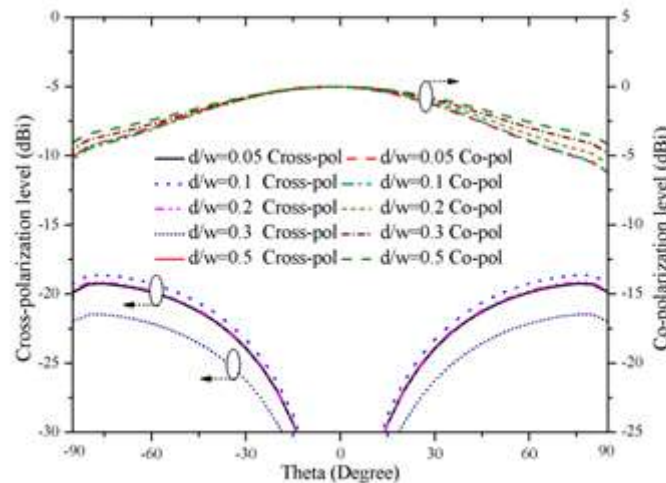


Fig. 6. Co-polarization in E-plane and cross-polarization in H-plane of the 50Ω microstrip line edge fed antenna with different d/w ratio

Due to use of 50Ω microstrip line feed cross-polarization level in H-plane is reduced to more than -20dB for all cases of d/w pin ratio as shown in Fig. 6, which is marked improvement as compared to results proposed in the literature [9].

IV. Result And Discussion

To validate the proposed design of radiating edge microstrip line fed configuration, two 50Ω microstrip edge fed RMSA with and without shorting pin are fabricated as shown in Fig. 7 and measurement are carried out using Anritsu vector network analyzer on substrate RO4003C having patch dimension of $50 \times 50\text{mm}^2$. The shorting post is located at $d/w=0.1$ with 50Ω microstrip line of $W_S=3.402\text{mm}$ and $L_S=10\text{mm}$. The ground plane dimension was $70 \times 70\text{mm}^2$. Fig. 8 depicts the surface current distribution of microstrip edge fed single shorted RMSA at corresponding

g_{FL} and F_F. All the simulation has been obtained by using the method of moments based software IE3D with infinite ground plane [10]. The simulated and measured results are compared in Figs. 9 and 10.

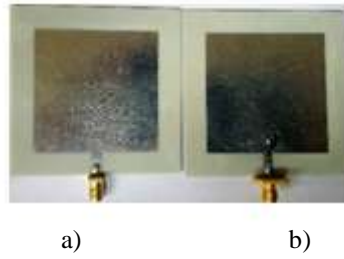


Fig. 7. Fabricated microstrip edge fed RMSA antenna (a) Without shorting pin and (b) With shorting Pin

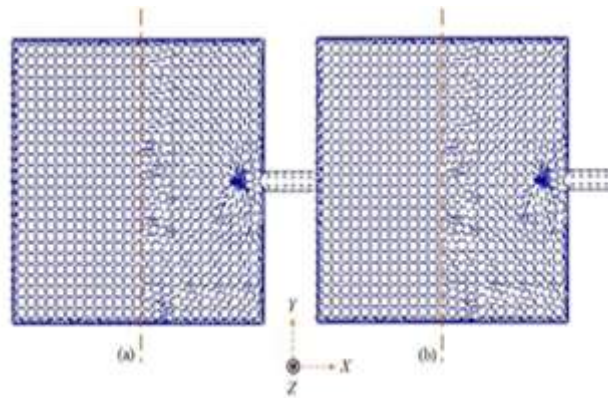


Fig. 8. Surface current distribution of microstrip edge fed single shorted RMSA at corresponding (a) F_L and (b) F_F

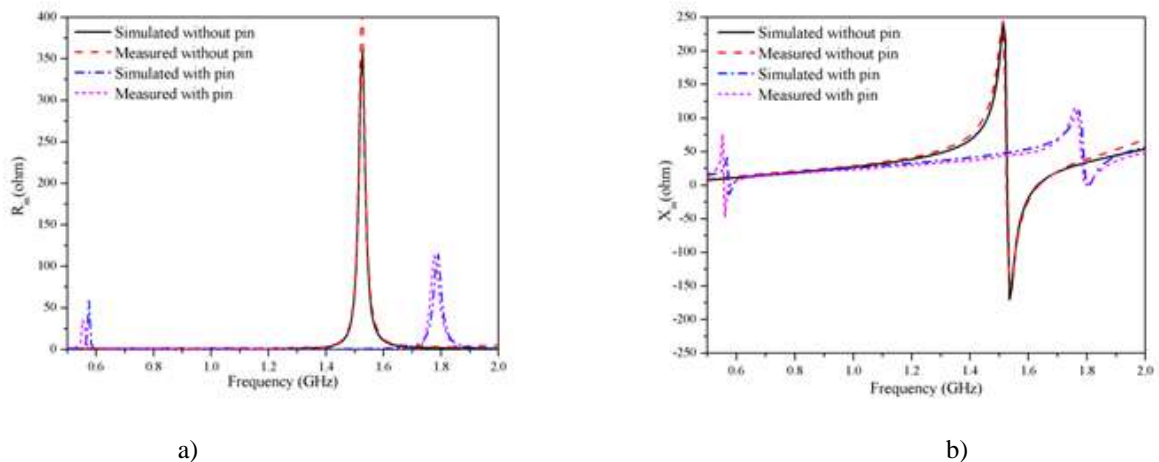


Fig. 9. Comparison of (a) Input resistance and (b) Input reactance of microstrip fed RMSA

Fig. 9(a) and (b) depict the simulated and measured results of input resistance and reactance of microstrip radiating edge fed RMSA. The input impedance of without and with shorting pin of the RMSA has reduced from 370 Ω to 50 Ω at corresponding F_F of 1.56 GHz and 1.8 GHz, respectively. Fig. 10 shows the simulated and measured radiation pattern of microstrip radiating edge fed RMSA with a shorting post at d/w=0.1 for F_F=1.8 GHz. The simulated and measured gain of 4.26 dBi and 3.15 dBi, respectively, is obtained at F_F=1.8 GHz and 1.802 GHz.

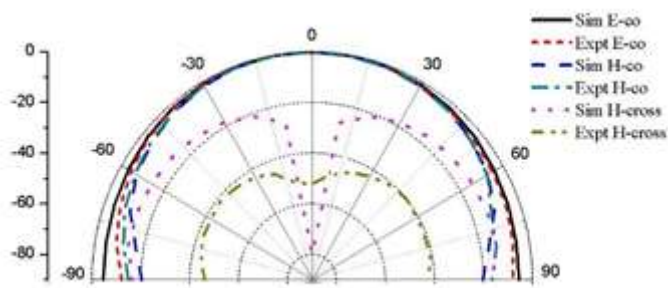


Fig. 10. Radiation pattern of microstrip edge fed RMSA with shorting pin at $d/w= 0.1$ for $F_F= 1.8$ GHz

V. Conclusion

In this paper, microstrip feed line of 50Ω impedance has been proposed to match the impedance of radiating edge fed RMSA using a shorting post. This method overcomes the disadvantages of using quarter wave transformer and inset feed technique. It has been shown that the radiating edge impedance of the MSAs can be adjusted in a wider range by selecting the shorting post location. An analysis of the lower and fundamental mode characteristics of shorted RMSAs are also presented. Simulation and measured results are in reasonable agreement.

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